**Preliminary** 



# SANYO Semiconductors DATA SHEET

# STK415-120-E 2-Channel Power Switching Audio Power IC, 80W+80W

#### Overview

The STK415-120-E is a class H audio power amplifier hybrid IC that features a built-in power supply switching circuit. This IC provides high efficiency audio power amplification by controlling (switching) the supply voltage supplied to the power devices according to the detected level of the input audio signal.

# **Applications**

• Audio power amplifiers.

#### **Features**

- Pin-to-pin compatible outputs ranging from 80W to 180W.
- Can be used to replace the STK416-100 series (3-channel models) and the class-AB series (2, 3-channel models) due to its pin compatibility.
- Pure complementary construction by new Darlington power transistors
- Output load impedance:  $R_L = 8\Omega$  to  $4\Omega$  supported
- Using insulated metal substrate that features superlative heat dissipation characteristics that are among the highest in the industry.

## **Series Models**

	STK415-090-E	STK415-100-E	STK415-120-E	STK415-130-E	STK415-140-E						
Output 1 (10%/1kHz)	80W×2ch	90W×2ch	120W×2ch	150W×2ch	180W×2ch						
Output 2 (0.8%/20Hz to 20kHz)	50W×2ch	60W×2ch	80W×2ch	100W×2ch	120W×2ch						
Max. rated V <sub>H</sub> (quiescent)	±60V	±65V	±73V	±80V	±80V						
Max. rated V <sub>L</sub> (quiescent)	±41V	±42V	±45V	±46V	±51V						
Recommended operating $V_H$ (8 $\Omega$ )	±37V	±39V	±46V	±51V	±52V						
Recommended operating V <sub>L</sub> (8Ω)	±27V	±29V	±32V	±34V	±32V						
Dimensions (excluding pin height)	64.0mm×31.1mm×9.0mm										

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# **Specifications**

Absolute maximum ratings at Ta=25°C (excluding rated temperature items), Tc=25°C unless otherwise specified

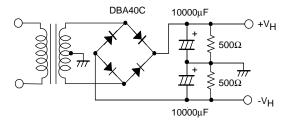
Parameter	Symbol	Conditions	Ratings	Unit
V <sub>H</sub> maximum quiescent supply voltage 1	V <sub>H</sub> max (1)	When no signal	±73	V
V <sub>H</sub> maximum supply voltage 2	V <sub>H</sub> max (2)	R <sub>L</sub> ≥6Ω	±65	V
V <sub>H</sub> maximum supply voltage 3	V <sub>H</sub> max (3)	R <sub>L</sub> ≥4Ω	±52	V
V <sub>L</sub> maximum quiescent supply voltage 1	V <sub>L</sub> max (1)	When no signal	±45	V
V <sub>L</sub> maximum supply voltage 2	V <sub>L</sub> max (2)	R <sub>L</sub> ≥6Ω	±42	V
V <sub>L</sub> maximum supply voltage 3	V <sub>L</sub> max (3)	R <sub>L</sub> ≥4Ω	±32	V
Maximum voltage between V <sub>H and</sub> V <sub>L</sub> *4	V <sub>H</sub> -V <sub>L</sub> max	No loading	60	V
Standby pin maximum voltage	Vst max		-0.3 to +5.5	V
Thermal resistance	θј-с	Per power transistor	1.6	°C/W
Junction temperature	Tj max	Both the Tj max and Tc max conditions must be met.	150	°C
IC substrate operating temperature	Tc max		125	°C
Storage temperature	Tstg		-30 to +125	°C
Allowable load shorted time *3	ts	$V_H$ =±46V, $V_L$ =±32V, $R_L$ =8 $\Omega$ , f=50Hz, $P_O$ =80W, 1-channel active	0.3	s

#### Electrical Characteristics at Tc=25°C, RL=8Ω (non-inductive load), Rg=600Ω, VG=40dB, VZ=15V

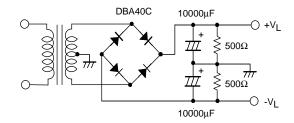
				Conc								
Parameter	Symbol		V (V)	f (Hz)	P <sub>O</sub> (W)	THD (%)		min	typ	max	unit	
Output power	P <sub>O</sub> (1)	V <sub>H</sub> V <sub>L</sub>	±46 ±32	20 to 20k		0.8		80				
	P <sub>O</sub> (2)	VH VL	±36 ±26	1k		0.8	R <sub>L</sub> =4Ω		80		W	
Total harmonic distortion	THD	VH VL	±46 ±32	20 to 20k	80				0.4		%	
Frequency characteristics	fL, fH	V <sub>H</sub> ±46 V <sub>L</sub> ±32			1.0	1.0 +0 -3dB			20 to 50k	(	Hz	
Input impedance	ri	V <sub>H</sub> V <sub>L</sub>	±46 ±32	1k	1.0				55		kΩ	
Output noise voltage *2	VNO	V <sub>H</sub> V <sub>L</sub>	±54 ±35				Rg=2.2kΩ			1.0	mVrms	
Quiescent current	Icco	٧ <sub>H</sub>	±54				R <sub>L</sub> =∞			30	mA	
		٧L	±35				KL=∞			100	IIIA	
Output neutral voltage	VN	∨ <sub>H</sub> ∨ <sub>L</sub>	±54 ±35					-70 0		+70	mV	
Pin 17 voltage when standby ON *7	VST ON	V <sub>H</sub> V <sub>L</sub>	±46 ±32				Standby		0	0.6	V	
Pin 17 voltage when standby OFF *7	VST OFF	V <sub>H</sub> V <sub>L</sub>	±46 ±32				Operating	2.5	3.0		V	

#### [Remarks]

- \*1: Unless otherwise specified, use a constant-voltage power supply to supply power when inspections are carried out.
- \*2: The output noise voltage values shown are peak values read with a VTVM. However, an AC stabilized (50Hz) power supply should be used to minimize the influence of AC primary side flicker noise on the reading.
- \*3: Use the designated transformer power supply circuit shown in the figure below for the measurements of allowable load shorted time and output noise voltage.
- \*4: Design circuits so that (VH-VI) is always less than 40V when switching the power supply with the load connected.
- \*5: Set up the VL power supply with an offset voltage at power supply switching (VL-VO) of about 8V as an initial target.
- \*6: Please connect –Pre V<sub>CC</sub> pin (#5 pin) with the stable minimum voltage and connect so that current does not flow in by reverse bias.
- \*7: Use the standby pin (pin 17) so that the applied voltage never exceeds the maximum rating. The power amplifier is turned on by applying +2.5V to +5.5V to the standby pin (pin 17).
- \*8: Thermal design must be implemented based on the conditions under which the customer's end products are expected to operate on the market.
- \*9: A thermoplastic adhesive resin is used for this hybrid IC.



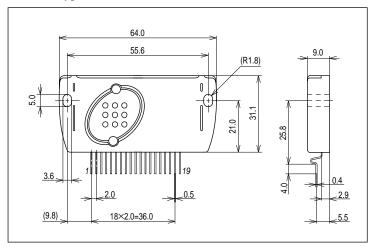
Designated transformer power supply (MG-250 equivalent)



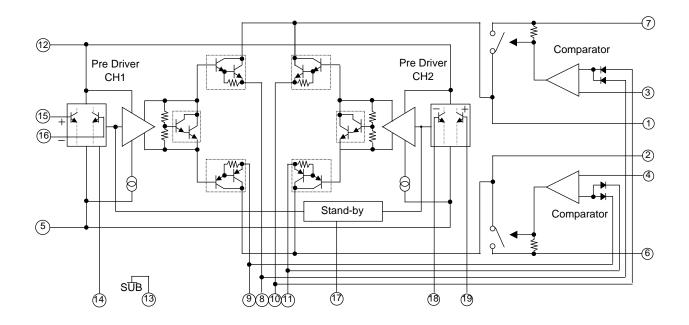
Designated transformer power supply (MG-200 equivalent)

# **Package Dimensions**

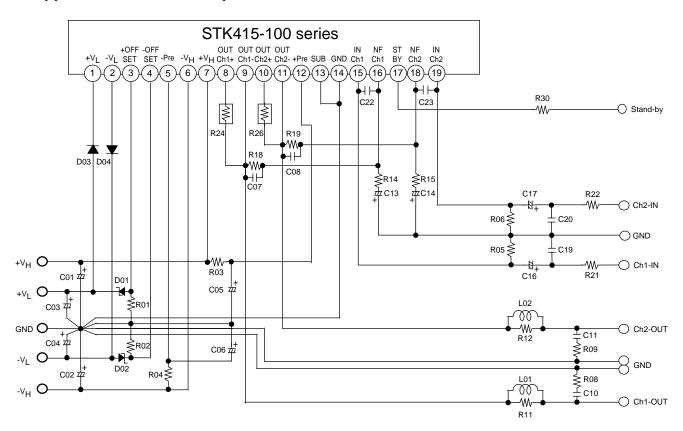
unit:mm (typ)



# **Internal Equivalent Circuit**



# **Application Circuit Example**



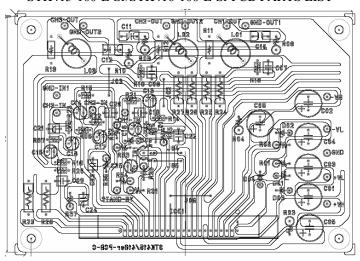


# **Recommended Values for Application Parts (for the test circuit)**

			, , , , , , , , , , , , , , , , , , , ,	1
Symbol	Recommended Value	Description	Larger than Recommended Value	Smaller than Recommended Value
R01, R02	1.5kΩ	Determine the current flowing into the power switching	Power holding circuit	Power switching circuit
,		circuit (comparator), (3mA to 10mA at V <sub>H</sub> power	remains active at lower	activates at higher
		switching)	frequencies.	frequencies.
R03, R04	100Ω/1W	Ripple filtering resistors	Decreased pass-through	Increased pass-through
1100, 1104	10022/177	(Used with C05 and C06 to form a ripple filter.)	current at high frequencies.	current at high frequencies.
R05, R06	56kΩ		VN offset	current at high frequencies.
K05, K06	30K22	Input bias resistors		when abonding \
D00 D00	4.70/4\\	(Virtually determine the input impedance.)	(Ensure R05=R18, R06=R19	when changing.)
R08, R09	4.7Ω/1W	Oscillation prevention resistor	-	-
R11, R12	4.7Ω	Oscillation prevention resistor	-	-
R14,R15	560Ω	Used with R18 and R19 to determine the voltage gain	Likely to oscillate	None
		VG. (VG should desirably be determined by the R14	(VG<40dB)	
		and R15 value.)		
R18, R19	56kΩ	Used with R14 and R15 to determine the voltage gain	-	-
		VG.		
R21, R22	1kΩ	Input filtering resistor	-	-
R24, R26	0.22Ω±10%,	Output emitter resistors	Decrease in maximum	Likely to cause thermal-
	5W	(Use of cement resistor is desirable)	output power	runaway.
R30	Remarks *7	Use a limiting resistor according to the voltage applied to	the standby pin so that it remain	ns within the rating.
C01, C02	100μF/	Oscillation prevention capacitors.		
	100V	Insert the capacitors as close to the IC as possible to		
		decrease the power impedance for reliable IC	-	-
		operation (use of electrolytic capacitors are		
		desirable).		
C03, C04	100μF/	Oscillation prevention capacitors.		
·	50V	Insert the capacitors as close to the IC as possible to		
		decrease the power impedance for reliable IC	-	-
		operation (use of electrolytic capacitors are		
		desirable).		
C05, C06	100μF/	Decoupling capacitors.	Increase in ripple component	s that pass into the input side
	100V	Eliminate ripple components that pass into the input	from the power line.	
		side from the power line.	·	
		(Used with R03 and R04 to form a ripple filter.)		
C07, C08	3pF	Oscillation prevention capacitor	Likely to oscillate	
C10, C11	0.1μF	Oscillation prevention capacitor	Likely to oscillate	
0.0,0	0	(Mylar capacitors are recommended.)	Zintoly to occurate	
C13, C14	22μF/	NF capacitor	Increase in low-frequency	Decrease in low-frequency
0.0, 0	10V	(Changes the low cutoff frequency;	voltage gain, with higher	voltage gain
	100	ex/f <sub>L</sub> =1/2π •C13•R14)	pop noise at power-on.	voltage gairi
046 047	2.2	_	pop notice at power on.	
C16, C17	2.2μF/	Input coupling capacitor (block DC current)	-	-
C40, C00	50V	lung t filter conscitor		
C19, C20	470pF	Input filter capacitor		
		(Used with R21 and R22 to form a filter that suppresses	-	-
000 000	100 =	high-frequency noises.)	19-1-69-4	<u> </u>
C22, C23	100pF	Oscillation prevention capacitor	Likely to oscillate.	Γ
D01, D02	15V	Determine the offset voltage at V <sub>L</sub> ↔V <sub>H</sub> power.	Decreased distortion at	Increased distortion at
			power switching time	power switching time.
D03, D04	3A/60V	Reverse current prevention diodes	_	-
	_	(FRD is recommended.)		
L01, L02	ЗμН	Oscillation prevention inductance	None	Likely to oscillate.

# **Sample PCB Trace Pattern**

 $STK415\text{-}100\text{-}E\text{-}Sr/STK416\text{-}100\text{-}E\text{-}Sr\ PCB\ PARTS\ LIST$ 



# **Parts List**

STK415, 416-100Sr PCB Parts List

PCB No.		PARTS	RATING	STK415 (416) -090-E, -100-E, -120-E, 130-E	STK415-140-E
R01, R02		-	ERX1SJ***	1.5kΩ, 1W	1.5kΩ, 1W
R03, R04		100Ω, 1W	ERG1SJ101	enabled	enabled
R05, R06, (R07), R18, R19, (R20)		56kΩ, 1/6W	RN16S563FK	enabled	enabled
R08, R09, (R10)		4.7Ω, 1W	ERX1SJ4R7	enabled	enabled
R11, R12, (R13)		4.7Ω, 1/4W	RN14S4R7FK	enabled	enabled
R14, R15, (R16)		-	RN16S***FK	560Ω, 1/6W	560Ω, 1/6W
R21, R22, (F	R23)	1kΩ, 1/6W	RN16S102FK	enabled	enabled
R25, R27, (F	R29)	0.22Ω±10%, 5W	BPR56CFR22J	Short	Short
R24, R26, (F	R28)	0.22Ω±10%, 5W	BPR56CFR22J	enabled	enabled
R35, R36, R	37	-	-	Short	Short
C01, C02, C05, C06		100μF, 100V	100MV100HC	enabled	enabled
C03, C04		100μF, 50V	50MV100HC	enabled	enabled
C07, C08, (C09)		3pF	DD104-63B3ROK50	enabled	enabled
C10, C11, (C12)		0.1μF, 100V	ECQ-V1H104JZ	enabled	enabled
C13, C14, (C15)		22μF, 10V	10MV220HC	enabled	enabled
C16, C17, (C18)		2.2μF, 50V	50MV2R2HC	enabled	enabled
C19, C20, (0	221)	470pF	DD104-63B471K50	enabled	enabled
C22, C23, (0	224)	100pF	DD104-63B101K50	enabled	enabled
D01, D02		-	-	GZA15X (SANYO)	GZA18X (SANYO)
D03, D04		IF (AV)=3A/60V		enabled	enabled
L01, L02, (L	03)	3μΗ		enabled	enabled
Stand-By	R30	3.3kΩ, 1/6W	RN16S332FK	enabled	enabled
	R32	1kΩ, 1/6W	RN16S102FK	enabled	enabled
	R33	33kΩ, 1/6W	RN16S333FK	enabled	enabled
	R34	2kΩ, 1/6W	RN16S202FK	enabled	enabled
	C25	47μF, 10V	10MV47HC	enabled	enabled
R33 R34		-	GMB01 (Ref.)	enabled	enabled
	TR1	-	2SC2274 (Ref.)	enabled	enabled
J01		Jumper	20mm	enabled	enabled
J02, J03, J0	6	Jumper	10mm	enabled	enabled
J04, J05		Jumper	7mm	enabled	enabled

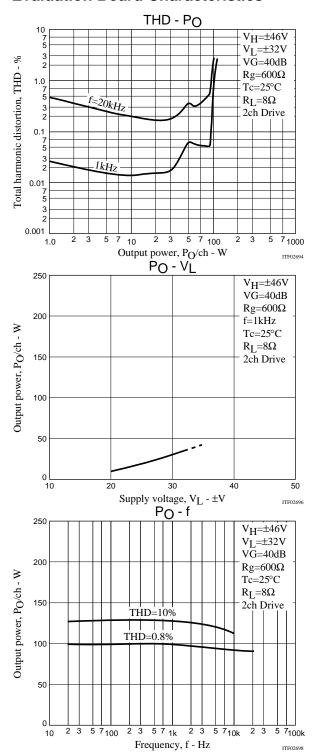
(\*1) STK416-100Sr (3ch AMP) doesn't mount parts of ().

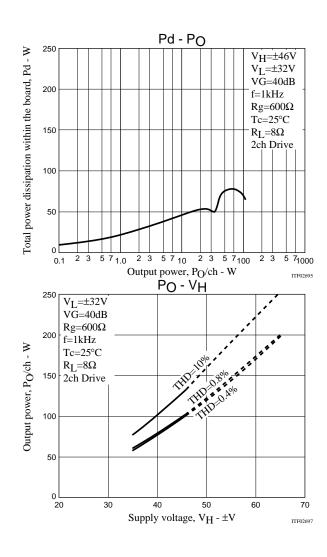
# **Pin Assignments**

[STK433-000/-100/-200 Sr & STK415/416-100 Sr Pin Layout]

[STK433-000/-100/-200 Sr & S	<u> </u>	+15/	410	-100	) Sr	Pin	Lay	out <sub>.</sub>															
2ch class-AB					1	2	3	4	5	6	7	8	9	10	11	12	13	14	15				
2011 Class-AD										2ch	clas	sAB/	2.00r	nm									
STK433-030-E 30W/JEITA					-	-	+	0	0	0	0	+			1	N	S	Ν	_				
STK433-040-E 40W/JEITA					Р	V	V	U	U	U	U	Р	S	G	Ν	F	Т	F	N				
STK433-060-E 50W/JEITA					R	С	С	Т	Т	Т	Т	R	U	Ν	/	/	Α	/	/				
STK433-070-E 60W/JEITA					Е	С	С	/	/	/	/	Е	В	D	С	С	Ν	С	С				
								С	С	С	С		•		Н	Н	D	Н	Н				
STK433-090-E 80W/JEITA								Н	Н	Н	Н		G		1	1	1	2	2				
STK433-100-E 100W/JEITA								1	1	2	2		Ν				В						
STK433-120-E 120W/JEITA								+	-	+	-		D				Υ						
STK433-130-E 150W/JEITA																							
3ch class-AB					1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19
סטוו טומסס־אט										3ch	clas	sAB/	2.00r	nm									
STK433-230A-E 30W/JEITA					-	-	+	0	0	0	0	+			1	N	S	Ν	1	I	Ν	0	0
STK433-240A-E 40W/JEITA					Р	٧	٧	U	U	U	U	Р	S	G	Ν	F	Т	F	N	N	F	U	U
STK433-260A-E 50W/JEITA					R	С	С	Т	Т	Т	Т	R	U	N	/	/	Α	/	/	/	/	Т	Т
STK433-270-E 60W/JEITA					Е	С	С	/	/	/	/	Е	В	D	С	С	Ν	С	С	С	С	/	/
STK433-290-E 80W/JEITA								С	С	С	С		•		Н	Н	D	Н	Н	Н	Н	С	С
STK433-300-E 100W/JEITA								Н	Н	Н	Н		G		1	1	-	2	2	3	3	Н	Н
STK433-320-E 120W/JEITA								1	1	2	2		N				В					3	3
STK433-330-E 150W/JEITA								+	-	+	-		D				Υ					+	-
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19				
2ch class-H										2c	h cla	ssH/2	2.00m	nm									
STK415-090-E 80W/JEITA	+	-	+	-	-	-	+	0	0	0	0	+			1	Ν	S	N	1				
STK415-100-E 90W/JEITA	V	٧	0	0	Р	٧	٧	U	U	U	U	Р	s	G	N	F	Т	F	N				
STK415-120-E 120W/JEITA	L	L	F	F	R	Н	Н	Т	Т	Т	Т	R	U	N	/	/	Α	/	/				
STK415-130-E 150W/JEITA			F	F	Е			/	/	/	/	Е	В	D	С	С	N	С	С				
STK415-140-E 180W/JEITA			s	s				С	С	С	С		•		Н	Н	D	Н	Н				
			Е	Е				Н	Н	Н	Н		G		1	1	ı	2	2				
			Т	Т				1	1	2	2		N				В						
								+	-	+	-		D				Υ						
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23
3ch class-H										3с	h cla	ssH/2	2.00m	ım									
STK416-090-E 80W/JEITA	+	-	+	-	-	-	+	0	0	0	0	+			I	Ν	S	N	I	I	Ν	0	0
STK416-100-E 90W/JEITA	V	٧	0	0	Р	٧	٧	U	U	U	U	Р	s	G	N	F	Т	F	N	N	F	U	U
STK416-120-E 120W/JEITA	L	L	F	F	R	Н	Н	Т	Т	Т	Т	R	U	N	/	/	Α	/	/	/	/	Т	Т
STK416-130-E 150W/JEITA			F	F	Е			/	/	/	/	Е	В	D	С	С	Ν	С	С	С	С	/	/
			s	s				С	С	С	С		•		Н	Н	D	Н	Н	Н	Н	С	С
			Е	Е				Н	Н	Н	Н		G		1	1		2	2	3	3	Н	Н
			Т	Т				1	1	2	2		N				В					3	3
								+	-	+	-		D				Υ					+	-

# **Evaluation Board Characteristics**





# **Discontinued**

#### STK415-120-E

[Thermal Design Example for STK415-120-E ( $R_L = 8\Omega$ )]

The thermal resistance,  $\theta$ c-a, of the heat sink for total power dissipation, Pd, within the hybrid IC is determined as follows

Condition 1: The hybrid IC substrate temperature, Tc, must not exceed 125°C.

$$Pd \times \theta c - a + Ta < 125^{\circ}C \qquad (1)$$

Ta: Guaranteed ambient temperature for the end product

Condition 2: The junction temperature, Tj, of each power transistor must not exceed 150°C.

$$Pd \times \theta c - a + Pd/N \times \theta j - c + Ta < 150^{\circ}C \qquad (2)$$

N: Number of power transistors

θj-c: Thermal resistance per power transistor

However, the power dissipation, Pd, for the power transistors shall be allocated equally among the number of power transistors.

The following inequalities result from solving equations (1) and (2) for  $\theta$ c-a.

$$\theta c-a < (125 - Ta)/Pd$$
 ..... (1)'  $\theta c-a < (150 - Ta)/Pd - \theta j-c/N$  .... (2)'

Values that satisfy these two inequalities at the same time represent the required heat sink thermal resistance.

When the following specifications have been stipulated, the required heat sink thermal resistance can be determined from formulas (1)' and (2)'.

Supply voltage
 Load resistance
 Guaranteed ambient temperature
 Ta

#### [Example]

When the IC supply voltage,  $V_H$ =±46V,  $V_L$ =±32V and  $R_L$  is  $8\Omega$ , the total power dissipation, Pd, within the hybrid IC, will be a maximum of 77W at 1kHz for a continuous sine wave signal according to the Pd-Po characteristics. For the music signals normally handled by audio amplifiers, a value of 1/8Po max is generally used for Pd as an estimate of the power dissipation based on the type of continuous signal. (Note that the factor used may differ depending on the safety standard used.)

This is:

Pd 
$$\approx 46.0$$
W (when 1/8PO max. = 10W, PO max. = 80W).

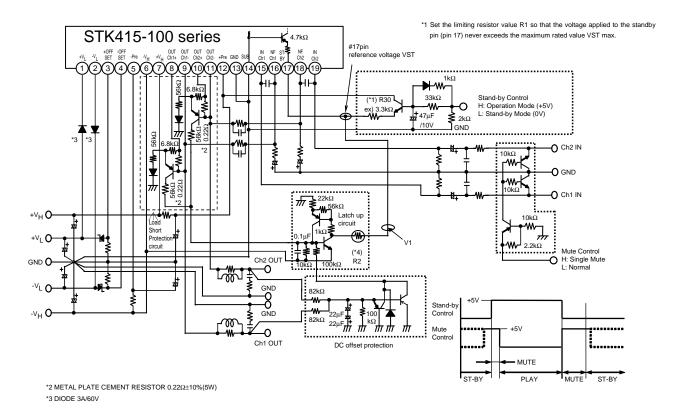
The number of power transistors in audio amplifier block of these hybrid ICs, N, is 4, and the thermal resistance per transistor,  $\theta$ j-c, is 1.6°C/W. Therefore, the required heat sink thermal resistance for a guaranteed ambient temperature, Ta, of 50°C will be as follows.

From formula (1)' 
$$\theta c\text{-}a < (125-50)/46.0 \\ < 1.63$$
 From formula (2)' 
$$\theta c\text{-}a < (150-50)/46.0 - 1.6/4 \\ < 1.77$$

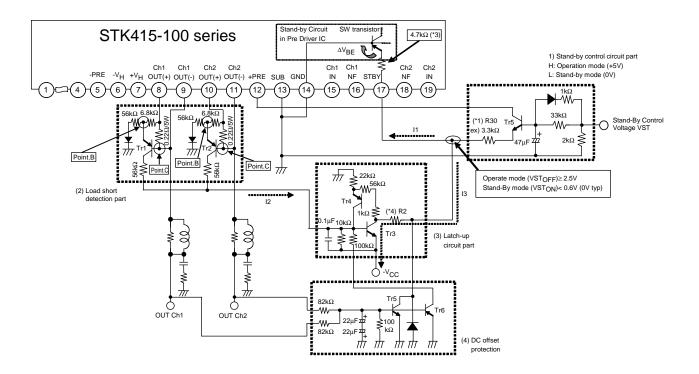
Therefore, the value of 1.63°C/W, which satisfies both of these formulae, is the required thermal resistance of the heat sink.

Note that this thermal design example assumes the use of a constant-voltage power supply, and is therefore not a verified design for any particular user's end product.

# STK415-100 Series Stand-by control, Mute control, Load-short protection & DC offset protection application



# STK415-100 Series Application explanation



The protection circuit application for the STK415-100sr consists of the following blocks (blocks (1) to (4)).

- (1) Standby control circuit block
- (2) Load short-circuit detection block
- (3) Latch-up circuit block
- (4) DC voltage protection block

#### 1) Standby control circuit block

Concerning pin 17 reference voltage VST

#### <1> Operation mode

The switching transistor of the predriver IC turns on when the pin 17 reference voltage, VST, becomes greater than or equal to 2.5V, placing the amplifier into the operation mode.

Example: When VST (min.) = 2.5V

I1 is approximately equal to 0.40mA since VST = (\*2) × IST + 0.6V  $\rightarrow$  2.5V = 4.7k $\Omega$  × IST + 0.6V.

#### <2> Standby mode

The switching transistor of the predriver IC turns off when the pin 17 reference voltage, VST, becomes lower than or equal to 0.6V (typ. 0V), placing the amplifier into the standby mode.

Example: When VST = 0.6V

I1 is approximately equal to 0mA since VST = (\*2) × IST + 0.6V  $\rightarrow$  0.6V = 4.7k $\Omega$  × IST + 0.6V.

#### (\*1) Limiting resistor

Determine the value of R1 so that the voltage VST applied to the standby pin (pin 17) falls within the rating (+2.5V to 5.5V (typ. 3.0V)).

- (\*2) The standby control voltage must be supplied from the host including microcontrollers.
- (\*3) A  $4.7k\Omega$  limiting resistor is also incorporated inside the hybrid IC (at pin 17).

#### 2) Load short-circuit detection block

Since the voltage between point B and point C is less than 0.6V in normal operation mode ( $V_{BE} < 0.6V$ ) and TR1 (or TR2) is not activated, the load short-circuit detection block does not operate.

When a load short-circuit occurs, however, the voltage between point B and point C becomes larger than 0.6V, causing TR1 (or TR2) to turn on  $(V_{BE} > 0.6V)$ , and current I2 to flows.

#### 3) Latch-up circuit block

TR3 is activated when I2 is supplied to the latch-up circuit.

When TR3 turns on and current I3 starts flowing, VST goes down to 0V (standby mode), protecting the power amplifier.

Since TR3 and TR4 configure a thyristor, once TR3 is activated, the IC is held in the standby mode.

To release the standby mode and reactivate the power amplifier, it is necessary to set the standby control voltage (\*2) temporarily low (0V). Subsequently, when the standby control is returned to high, the power amplifier will become active again.

(\*4) The I3 value varies depending on the supply voltage. Determine the value of R2 using the formula below, so that I1 is equal to or less than I3.

 $I1 \le I3 = V_{CC}/R2$ 

# 4) DC offset protection block

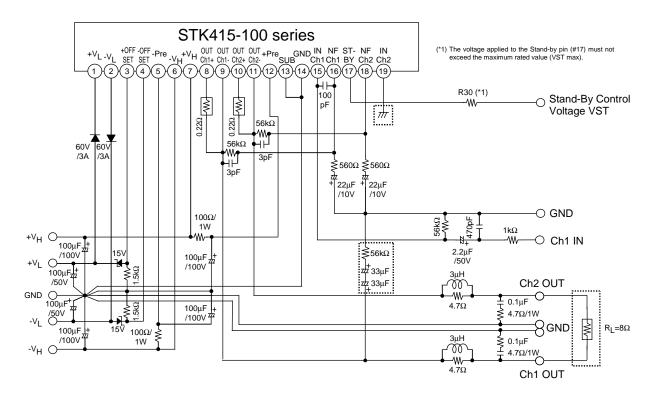
The DC offset protection circuit is activated when  $\pm 0.5$ V (typ) voltage is applied to either "OUT CH1" or "OUT CH2," and the hybrid IC is shut down (standby mode).

To release the IC from the standby mode and reactivate the power amplifier, it is necessary to set the standby control voltage temporarily low (0V).

Subsequently, when the standby control is returned to high (+5V, for example), the power amplifier will become active again.

The protection level must be set using the  $82k\Omega$  resistor. Furthermore, the time constant must be determined using  $22\mu//22\mu$  capacitors to prevent the amplifier from malfunctioning due to the audio signal.

# STK415-100 Series BTL Application



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